

Optimizing the Energy-Delay-Ringing Product in On-Chip CMOS Line Drivers

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Outline

- Introduction
- Problem Statement
- Transmission Line Modeling
- Energy-Delay-Ringing Product(New Metric)
- Simulation Results
- Summary

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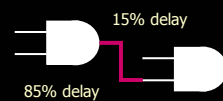
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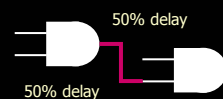
Introduction

- System-level electrical issues are becoming more critical
 - Higher clock rates
 - wires are transmission lines
 - clock skew and jitter are a major portion of a clock cycle
 - Lower voltages
 - more current for a given power level
 - less margin
 - Pin bottlenecks
 - need to make each signal count

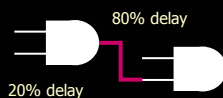
Wire Delay



- Mid 80 Scenario
 - Most of the input to output delay of the logic is due to gate delay

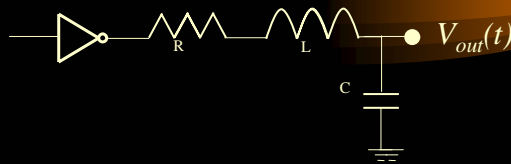


- Mid 90 Scenario
 - Half of the input to output delay of the logic is due to wire delay

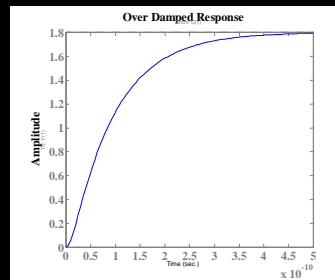


- Today's Scenario
 - Most of the input to output delay of the logic is due to wire delay

Problem Statement



Weak inductive effect

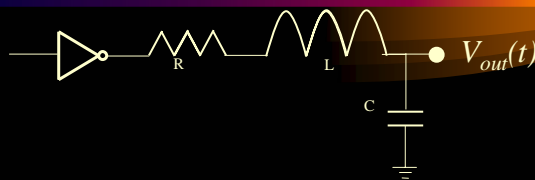


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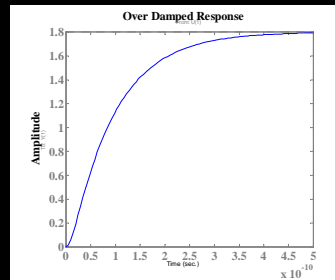
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Problem Statement



Significant inductive effect

Weak inductive effect

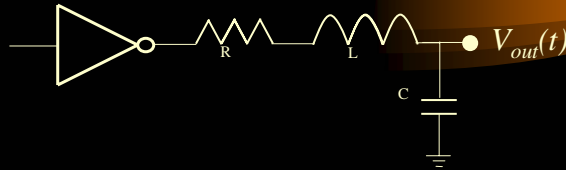


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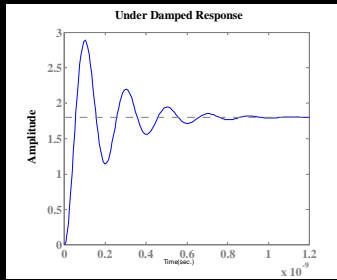
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Problem Statement

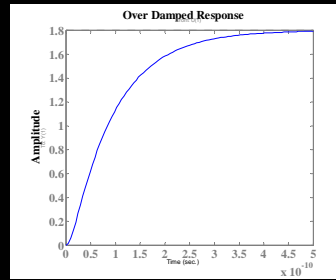


Significant inductive effect



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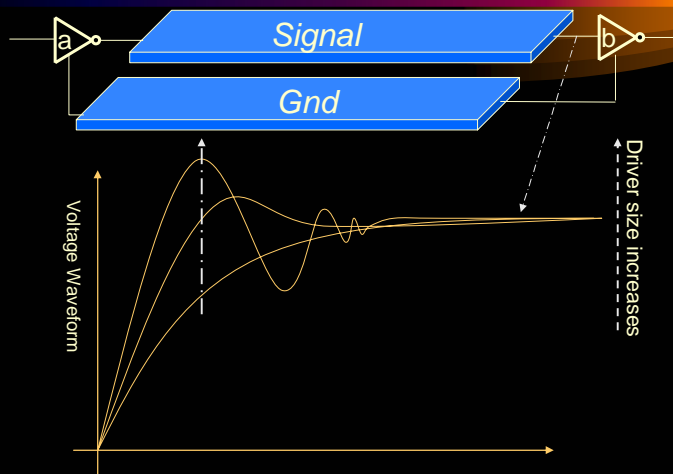
Weak inductive effect



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Energy, Delay, and Ringing

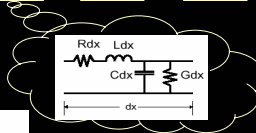
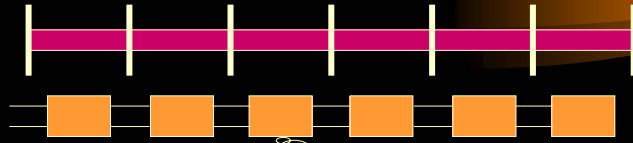


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Transmission Line Theory



Time Domain		Frequency Domain
$\nabla \cdot \epsilon E = \frac{\rho}{\epsilon}$		$\epsilon E = \frac{\rho}{\epsilon_0}$
$\nabla \cdot B = 0$		$\mu H = 0$
$\nabla \times E + \frac{\partial B}{\partial t} = 0$		$+ j\omega \mu H = 0$
$\nabla \times B - \frac{\mu \epsilon \partial E}{c^2 \partial t} = J$		$- j\omega \epsilon E = J$

+ Boundary Conditions

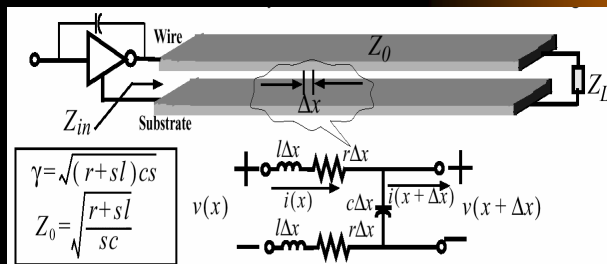
- We should think of Maxwell equations to analyze the transmission lines accurately.

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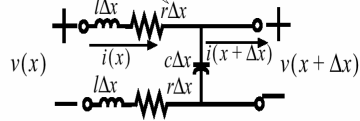
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Characteristics of the Line



$$\gamma = \sqrt{(r+sl)cs}$$

$$Z_0 = \sqrt{\frac{r+sl}{sc}}$$



$$Z_{in}|_{x=-h} = Z_0 \frac{Z_L + Z_0 \tanh(\gamma h)}{Z_0 + Z_L \tanh(\gamma h)}$$

$$H_v|_{x=-h} = \frac{Z_L}{Z_0 \cosh(\gamma h) + Z_L \sinh(\gamma h)}$$

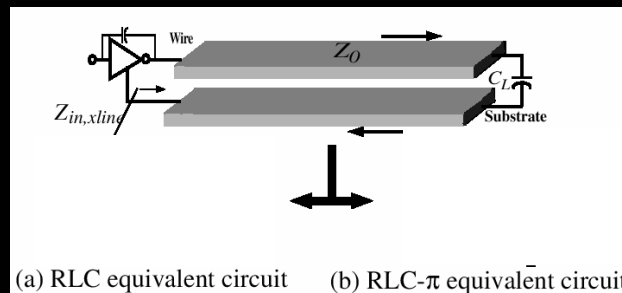
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New Energy Model

- Need to efficiently model the energy dissipation of on-chip interconnect driven by CMOS drivers
 - Truncating the $Z_{in}(s)$ gives you a model for energy



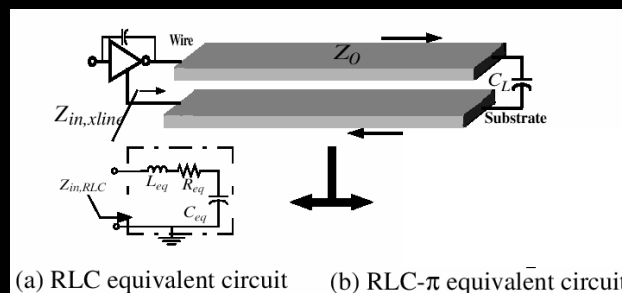
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New Energy Model

- Need to efficiently model the energy dissipation of on-chip interconnect driven by CMOS drivers
 - Truncating the $Z_{in}(s)$ gives you a model for energy
 - First order truncation: $\tanh(\gamma h) \rightarrow \gamma h$ gives RLC model



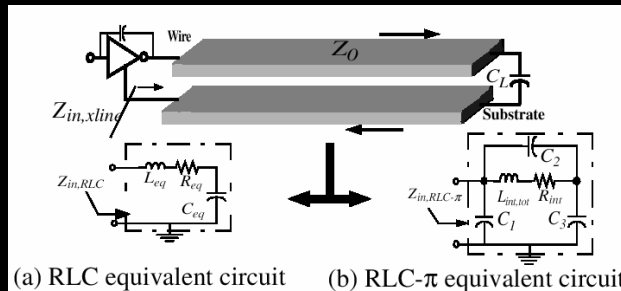
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New Energy Model

- Need to efficiently model the energy dissipation of on-chip interconnect driven by CMOS drivers
 - Truncating the $Z_{in}(s)$ gives you a model for energy
 - First order truncation: $\tanh(\gamma h) \rightarrow \gamma h$ gives RLC model
 - Second order truncation: $\tanh(\gamma h) \rightarrow 2\gamma h / (2 + \gamma^2 h^2)$ results in an RLC- π model.

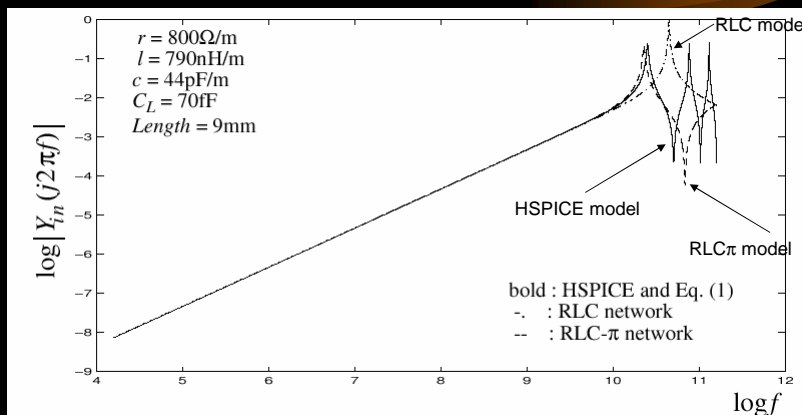


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RLC- π vs. RLC Energy Model



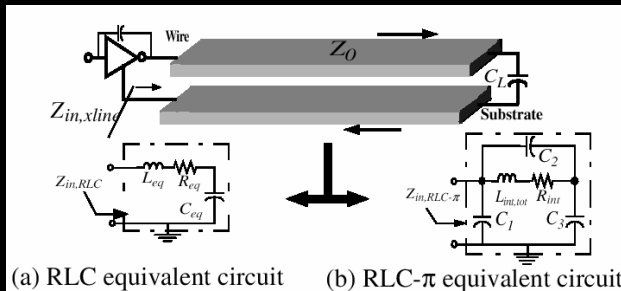
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New Delay Model

- Need to efficiently model the delay of the on-chip interconnect
 - Truncating the $H(s)$ results in a model for delay
 - The same scenario holds for delay with new values of R , L , C_1 , C_2 and C_3 .

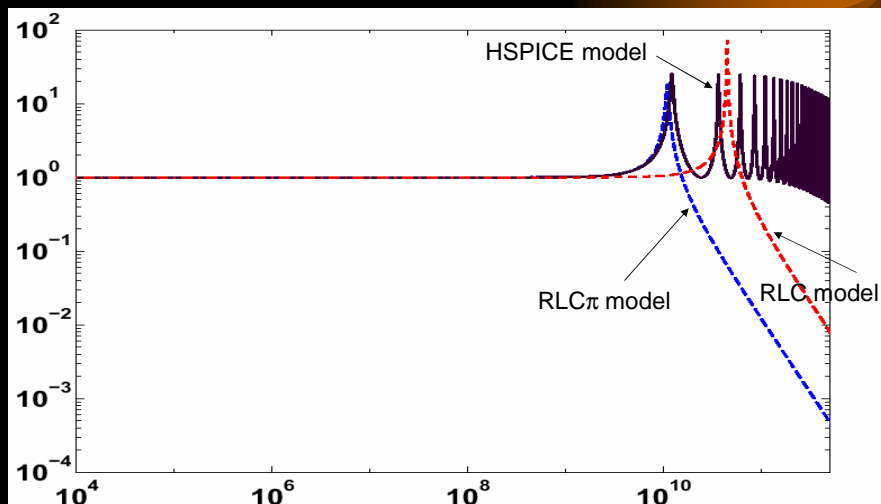


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RLC vs. RLC- π Delay Model



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RLC- π Energy and Delay Model

- Energy Parameters of Transmission Line

$$C_3^E = \sqrt{\frac{(C_{int,tot} + C_L)^2 + C_L^2}{2}}, \quad C_2^E = \frac{C_{int,tot}}{2} + C_L - C_3^E$$

$$C_1^E = C_{int,tot} + C_L - C_3^E$$

- Delay Parameters of Transmission Line

$$C_1^D = \frac{C_{int,tot}}{2}, \quad C_2^D = 0, \quad C_3^D = \frac{C_{int,tot}}{2} + C_L$$

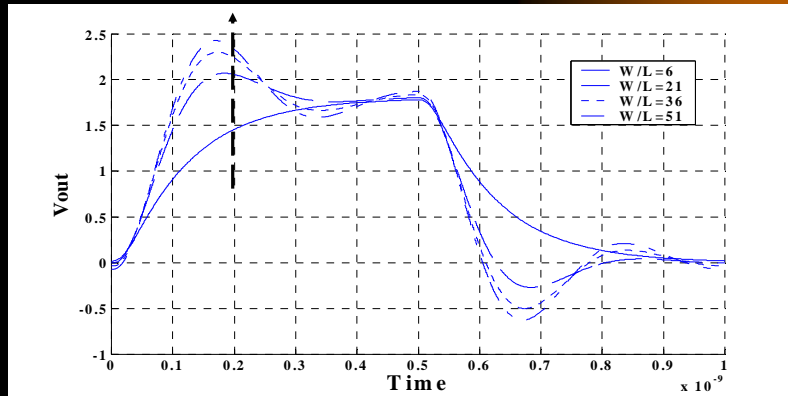
$$L_{eq}^D = L_{int,tot}, \quad R_{eq}^D = R_{int,tot}$$

Energy-Delay-Ringing Product



- The wire length is determined by the routing constraints.
- Driver upsizing
 - will increase the energy dissipation
 - will decrease the 50% propagation delay
- Increasing the current drive capability of the line driver causes the inductive effects to become more important
 - Ringing will increase the settling time
- Change the driver size to achieve the best Energy-Delay-Ringing Product

Output Voltages for Different Driver Sizes



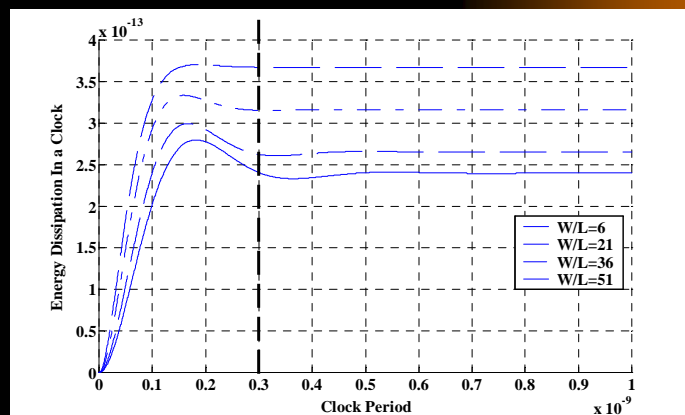
- V_{out} as a function of four different W/L driver values

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Energy for Different Driver Sizes



- Energy dissipation in a clock cycle as a function of the clock period for four different driver W/L ratios

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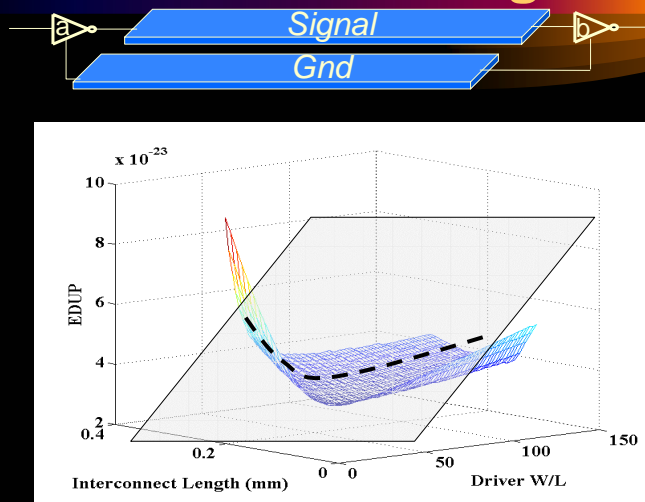
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Why EDR-P

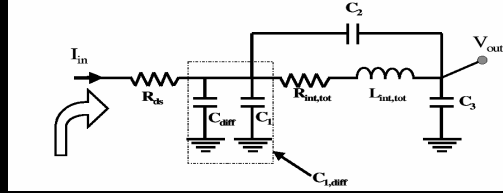
- Overdamped response \longleftrightarrow 50% delay is important
- Underdamped response \longleftrightarrow settling time is important
- Changing the driver size, varies the output waveform from overdamped to underdamped behavior.
- Propose a new metric; *Energy-Delay-Ringing Product*

EDR Product Diagram

The size of driver "p" is constant.



Energy and Delay Analysis



$$\frac{V_{out}(s)}{V_{in}(s)} = \frac{a_2 s^2 + a_1 s + a_0}{b_3 s^3 + b_2 s^2 + b_1 s + b_0}$$

where

$$\begin{aligned} a_2 &= L_{int} C_2^E & a_1 &= R_{int} C_2^E & a_0 &= 1 \\ b_3 &= R_{ds} L_{int} (C_{1,diff}^E C_2^E + C_{1,diff}^E C_3^E + C_2^E C_3^E) \\ b_2 &= L_{int} (C_2^E + C_3^E) + R_{ds} R_{int} (C_{1,diff}^E C_2^E + C_{1,diff}^E C_3^E + C_2^E C_3^E) \\ b_1 &= R_{ds} (C_{1,diff}^E + C_3^E) + R_{int} (C_2^E + C_3^E) & b_0 &= 1 \end{aligned}$$

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Approximation

$$\frac{V_{out}(s)}{V_{in}(s)} = \frac{a_2 s^2 + a_1 s + a_0}{b_3 s^3 + b_2 s^2 + b_1 s + b_0}$$



$$\frac{V_{out}(s)}{V_{in}(s)} \cong \frac{a_2 s^2 + a_1 s + a_0}{b_2 s^2 + b_1 s + b_0}$$

$$\frac{I_{in}(s)}{V_{in}(s)} = \frac{d_3 s^3 + d_2 s^2 + d_1 s}{b_3 s^3 + b_2 s^2 + b_1 s + b_0} = 1 - \frac{L_{int} (C_2^E + C_3^E) s^2 + R_{int} (C_2^E + C_3^E) s + 1}{L_{int} C_2^E s^2 + R_{int} C_2^E s + 1} \times \frac{V_{out}(s)}{V_{in}(s)}$$

$$\begin{aligned} d_3 &= L_{int} (C_{1,diff}^E C_2^E + C_{1,diff}^E C_3^E + C_2^E C_3^E) \\ d_2 &= R_{int} (C_{1,diff}^E C_2^E + C_{1,diff}^E C_3^E + C_2^E C_3^E) \\ d_1 &= C_{1,diff}^E + C_3^E \end{aligned}$$

$$\frac{I_{in}(s)}{V_{in}(s)} = \frac{d_2 s^2 + d_1 s}{b_2 s^2 + b_1 s + b_0}$$

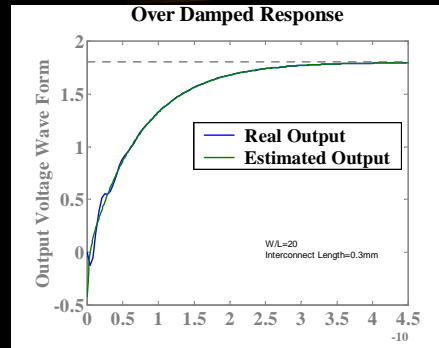
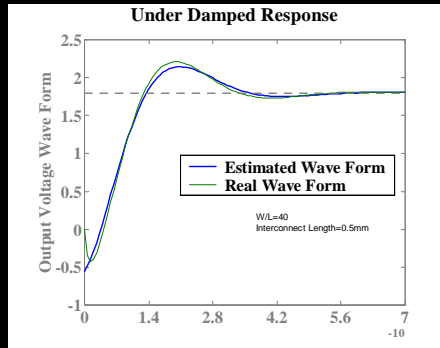
The same scenario holds for delay analysis

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Comparison for Two Different Responses

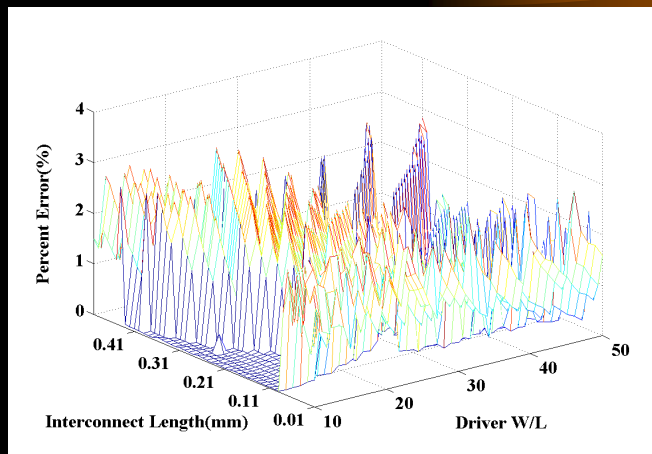


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Approximation Error

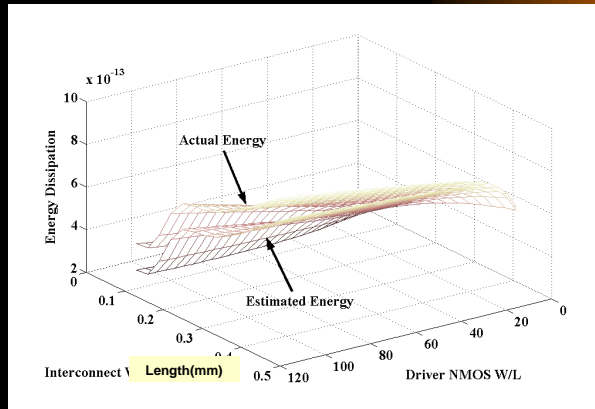


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Energy Dissipation: Simulation vs. Analytical Model



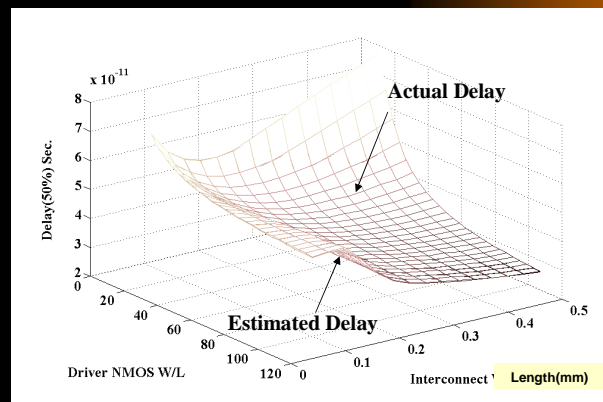
- Difference between simulation and analytical model for different interconnect lengths and driver sizes

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Delay: Simulation vs. Analytical Model



- Difference between real Delay and estimated Delay in 0.18 μ m technology for different interconnect lengths and driver sizes

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Summary



- We presented new transmission line energy and delay models for high frequencies
- We defined a new metric which is EDR-P
- We proposed closed form expressions for energy, delay and ringing.
- Simulation results verify the accuracy of the proposed analytical model.